

ECEN 314: Signals and Systems

Lecture Notes 5: Properties of DT Convolution

Reading:

- Current: SSOW 2.3
- Next: SSOW 2.4

1 Properties of Convolution

Property 1 (The sifting property). *Convolution with a time-shifted impulse results in a time-shifted version of the output:*

$$x[n] * \delta[n - n_0] = x[n - n_0].$$

Therefore, $x[n] * \delta[n] = x[n]$ and $\delta[n]$ is the identity element under convolution.

Proof. By definition,

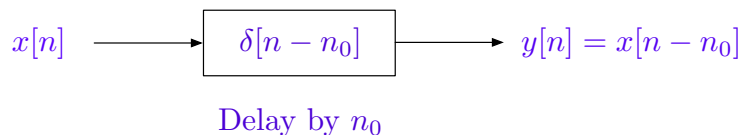
$$x[n] * \delta[n - n_0] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - k - n_0]$$

and $x[k] \delta[n - k - n_0] = x[n - n_0] \delta[n - k - n_0]$ because

$$\delta[n - k - n_0] = \begin{cases} 1, & k = n - n_0 \\ 0, & k \neq n - n_0 \end{cases}$$

Computing the sum shows that $y[n] = x[n - n_0]$ and therefore $x[n] * \delta[n - n_0] = x[n - n_0]$. \square

We can interpret this property using the following block diagram:



Property 2 (The commutative property). *The convolution operator is commutative:*

$$y[n] = x[n] * h[n] = h[n] * x[n].$$

Proof. By definition,

$$x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

and

$$h[n] * x[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$$

For the first, sum we can substitute $k = n - k'$ to get

$$x[n] * h[n] = \sum_{k'=-\infty}^{\infty} x[n-k']h[k'],$$

where the limits of summation are unchanged because they are infinite. We conclude that $x[n] * h[n] = h[n] * x[n]$. □

Property 3 (The distributive property). *Convolution distributes over addition:*

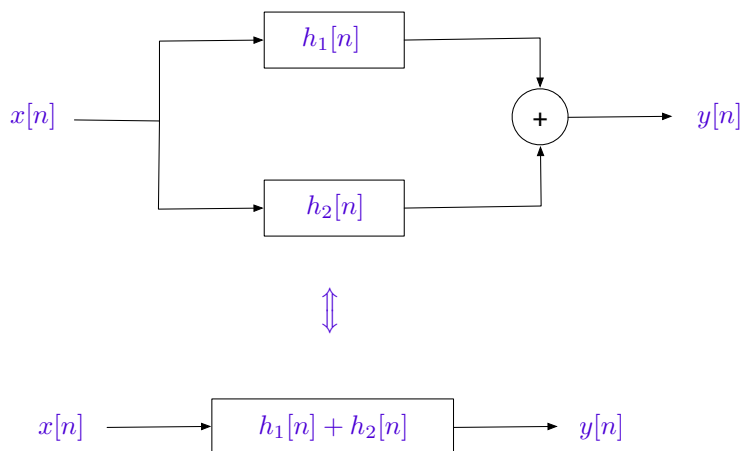
$$x[n] * (h_1[n] + h_2[n]) = x[n] * h_1[n] + x[n] * h_2[n].$$

Proof. By definition,

$$\begin{aligned} x[n] * (h_1[n] + h_2[n]) &= \sum_{k=-\infty}^{\infty} x[k](h_1[n-k] + h_2[n-k]) \\ &= \sum_{k=-\infty}^{\infty} x[k]h_1[n-k] + \sum_{k=-\infty}^{\infty} x[k]h_2[n-k] \\ &= x[n] * h_1[n] + x[n] * h_2[n] \end{aligned}$$

□

Interpretation:



Property 4 (The associative property). $x[n] * (h_1[n] * h_2[n]) = (x[n] * h_1[n]) * h_2[n]$

Proof. Let

$$a[n] = h_1[n] * h_2[n] = \sum_{k=-\infty}^{\infty} h_1[k]h_2[n-k]$$

and

$$b[n] = x[n] * h_1[n] = \sum_{l=-\infty}^{\infty} x[l]h_1[n-l]$$

Then

$$\begin{aligned} x[n] * a[n] &= \sum_{l=-\infty}^{\infty} x[l]a[n-l] \\ &= \sum_{l=-\infty}^{\infty} x[l] \left(\sum_{k=-\infty}^{\infty} h_1[k]h_2[n-l-k] \right) \end{aligned}$$

and

$$\begin{aligned} b[n] * h_2[n] &= \sum_{k=-\infty}^{\infty} b[k]h_2[n-k] \\ &= \sum_{k=-\infty}^{\infty} \left(\sum_{l=-\infty}^{\infty} x[l]h_1[k-l] \right) h_2[n-k] \\ &= \sum_{l=-\infty}^{\infty} x[l] \left(\sum_{k=-\infty}^{\infty} h_1[k-l]h_2[n-k] \right) \end{aligned}$$

Let $k' = k - l$. We have $k = l + k'$ and hence

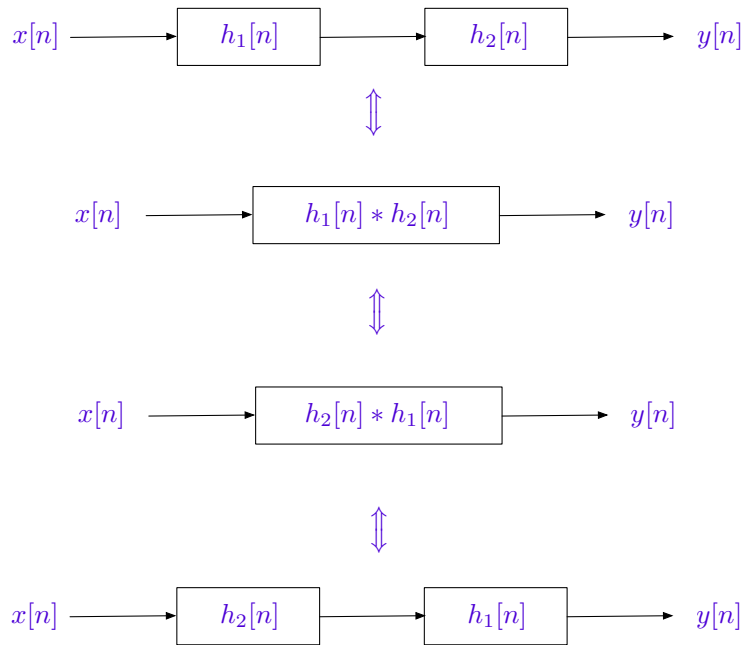
$$b[n] * h_2[n] = \sum_{l=-\infty}^{\infty} x[l] \left(\sum_{k'=-\infty}^{\infty} h_1[k']h_2[n-l-k'] \right)$$

Note that k and k' are indices of summation and, therefore, we conclude that $x[n] * a[n] = b[n] * h_2[n]$. \square

Combining properties 2 and 4, we have

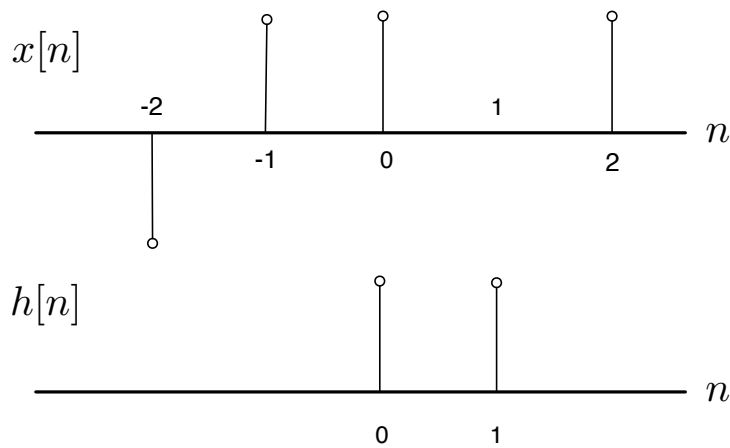
$$(x[n] * h_1[n]) * h_2[n] = x[n] * (h_1[n] * h_2[n]) = x[n] * (h_2[n] * h_1[n]) = (x[n] * h_2[n]) * h_1[n]$$

Interpretation:



Combining properties 1 and 3, one can compute the convolution of two finite signals algebraically.

Example 1. Consider the signals



and observe that

$$x[n] = -\delta[n + 2] + \delta[n + 1] + \delta[n] + \delta[n - 2]$$

and

$$h[n] = \delta[n] + \delta[n - 1].$$

Thus,

$$\begin{aligned}
 x[n] * h[n] &= (-\delta[n+2] + \delta[n+1] + \delta[n] + \delta[n-2]) * (\delta[n] + \delta[n-1]) \\
 &= -\delta[n+2] + \delta[n+1] + \delta[n] + \delta[n-2] + \\
 &\quad (-\delta[n+1] + \delta[n] + \delta[n-1] + \delta[n-3]) \\
 &= -\delta[n+2] + 2\delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3]
 \end{aligned}$$

Property 5 (Causality). *A DT LTI system is causal if and only if its unit impulse response $h[n] = 0$ for all $n < 0$.*

Proof. If a DT LTI system is causal, then the unit impulse response $h[n] = 0$ for all $n < 0$, as its input $\delta[n] = 0$ for all $n < 0$. Conversely, if the unit impulse response $h[n] = 0$ for all $n < 0$, by the convolution sum

$$y[n] = x[n] * h[n] = h[n] * x[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k] = \sum_{k \geq 0} h[k]x[n-k]$$

i.e., $y[n]$ is fully determined by the current and previous inputs $x[n-k]$ for $k \geq 0$. We conclude that the system must be causal. \square

Property 6 (Stability). *A DT LTI system is stable if and only if $\sum_{k=-\infty}^{\infty} |h[k]| < \infty$.*

Proof. If $A = \sum_{k=-\infty}^{\infty} |h[k]| < \infty$ and the input $x[n]$ is bounded (i.e., $|x[n]| \leq M_x < \infty$ for all n), then

$$|y[n]| = \left| \sum_{k=-\infty}^{\infty} x[k]h[n-k] \right| \leq \sum_{k=-\infty}^{\infty} |x[k]||h[n-k]| \leq M_x \sum_{k=-\infty}^{\infty} |h[n-k]| = M_x A < \infty.$$

The proof of the "only if" part is outlined in SSOW problem 2.49. \square

A final word: We said all along that a DT LTI system is fully characterized by its unit impulse response $h[n]$. However, if a system is not LTI to begin with, knowing the unit impulse response $h[n]$ may shed very little insight about what the system is. In fact, many systems may share the same unit impulse response $h[n]$. Consider, for example, $h[n] = \delta[n]$ is the unit impulse response to any system $y[n] = (x[n])^k$ with an integer k . But among them, there is only one system that is LTI. That is, $y[n] = x[n]$.